

Overview of Models for Automated Process Control

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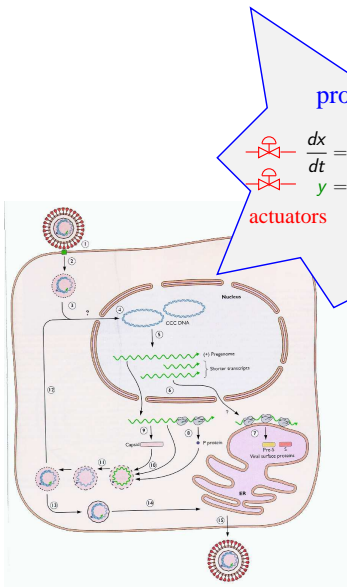


April 29, 2009

Utilization of Process Modeling and Advanced Process Control
in QbD based Drug Development and Manufacturing Workshop
Baltimore, MD

- 1 Introduction and Motivation
- 2 Modeling
 - Nonlinear differential equations
 - Linear time invariant models
 - Discrete time models
 - Input-output models
 - Constraints
 - Modeling the noise
- 3 From Models to Automated Process Control
 - Regulation
 - State estimation
 - Disturbance modeling and zero offset
- 4 Future Developments
- 5 Further Reading

The power of abstraction



process



actuators

$$\frac{dx}{dt} = f(x, u)$$

$$y = g(x, u)$$



sensors



Large industrial success story!

Linear MPC and ethylene manufacturing

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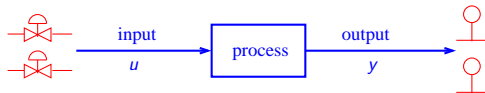
Praxair experience with MPC

- Praxair currently has more than 150 MPC installations
- 16 M\$/year increased profit (2008)

Broader industrial impact (Qin and Badgwell, 2003)

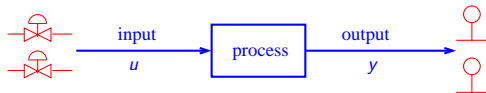
Area	Aspen Technology	Honeywell Hi-Spec	Adersa	PCL	MDC	Total
Refining	1200	480	280	25		1985
Petrochemicals	450	80	-	20		550
Chemicals	100	20	3	21		144
Pulp and Paper	18	50	-	-		68
Air & Gas	-	10	-	-		10
Utility	-	10	-	4		14
Mining/Metallurgy	8	6	7	16		37
Food Processing	-	-	41	10		51
Polymer	17	-	-	-		17
Furnaces	-	-	42	3		45
Aerospace/Defense	-	-	13	-		13
Automotive	-	-	7	-		7
Unclassified	40	40	1045	26	450	1601
Total	1833	696	1438	125	450	4542
First App.	DMC:1985 IDCOM-M:1987 OPC:1987	PCT:1984 RMPCT:1991	IDCOM:1973 HIECON:1986	PCL: 1984	SMOC: 1988	
Largest App	603x283	225x85	-	31x12	-	

Input, output, and state variables

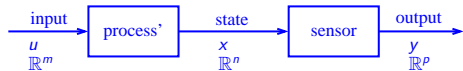


Input/output description

Input, output, and state variables



Input/output description



State description

Nonlinear differential equations

$$\frac{dx}{dt} = f(x, u, t)$$

$$y = h(x, u, t)$$

$$x(t_0) = x_0$$

- x is the state
- u is the input
- y is the output
- t is time

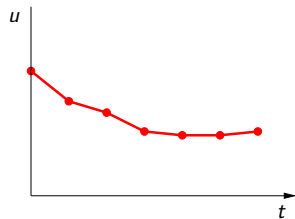
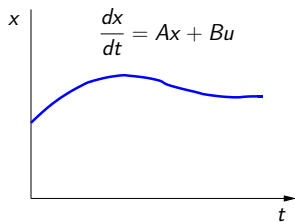
Linear differential equations

$$\frac{dx}{dt} = Ax + Bu$$

$$y = Cx + Du$$

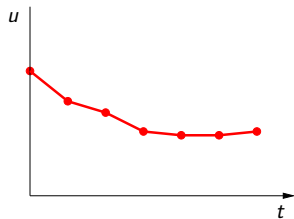
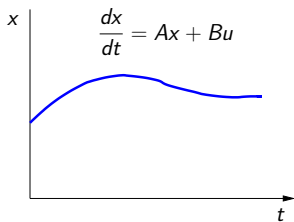
$$x(0) = x_0$$

Continuous time

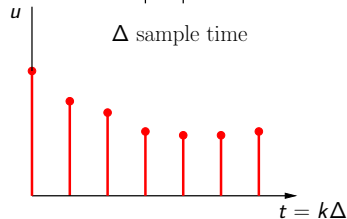
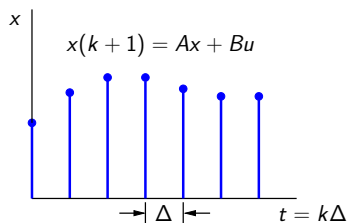


Continuous time and discrete time models

Continuous time



Discrete time



Linear difference equations

$$x(k + 1) = Ax(k) + Bu(k)$$

$$y(k) = Cx(k) + Du(k)$$

$$x(0) = x_0$$

Linear difference equations

$$\begin{aligned}x(k+1) &= Ax(k) + Bu(k) \\y(k) &= Cx(k) + Du(k) \\x(0) &= x_0\end{aligned}$$

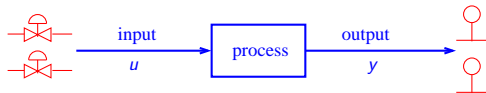
Easier notation

Linear difference equations

$$\begin{aligned}x^+ &= Ax + Bu \\y &= Cx + Du \\x(0) &= x_0\end{aligned}$$

in which $x(k)^+$ denotes $x(k+1)$.

Input-output models



Input/output description



State description

- Current output $y(t)$ is modeled as a time series in past *inputs* and *outputs*, and noise

$$y(t) = \underbrace{a_1 y(t-1) + a_2 y(t-2)}_{\text{past outputs}} + \underbrace{b_1 u(t-1) + b_2 u(t-2)}_{\text{past inputs}} + v(t)$$

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$$\theta = [a_1 \quad a_2 \quad b_1 \quad b_2]$$

- Given $\hat{\theta}$ we can obtain A, B, C

Converting input-output to state space

- Define $x(t)$ to hold the past inputs and outputs.

$$x(t) = \begin{bmatrix} y(t) \\ y(t-1) \\ u(t-1) \end{bmatrix}$$

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- The time series model is

$$\underbrace{\begin{bmatrix} y(t+1) \\ y(t) \\ u(t) \end{bmatrix}}_{x^+} = \underbrace{\begin{bmatrix} a_1 & a_2 & b_2 \\ 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix}}_A \underbrace{\begin{bmatrix} y(t) \\ y(t-1) \\ u(t-1) \end{bmatrix}}_x + \underbrace{\begin{bmatrix} b_1 \\ 0 \\ 1 \end{bmatrix}}_B u(t)$$

$$y(t) = \underbrace{[1 \quad 0 \quad 0]}_C \underbrace{\begin{bmatrix} y(t) \\ y(t-1) \\ u(t-1) \end{bmatrix}}_x$$

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- The equivalent state space model is

$$x^+ = Ax + Bu$$

$$y = Cx$$

Constraints

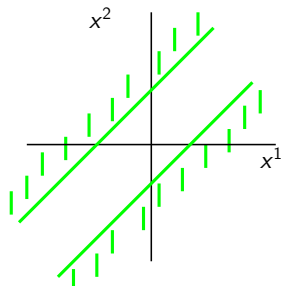
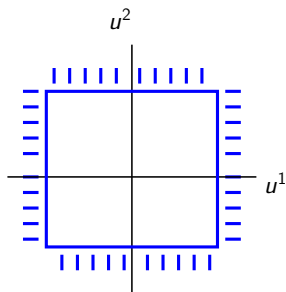
- Physical bounds on inputs (valves full open, full closed)

$$\underline{u} \leq u(k) \leq \bar{u} \quad k \geq 0$$

$$Eu(k) \leq e \quad k \geq 0$$

in which

$$E = \begin{bmatrix} I \\ -I \end{bmatrix} \quad e = \begin{bmatrix} \bar{u} \\ -\underline{u} \end{bmatrix}$$



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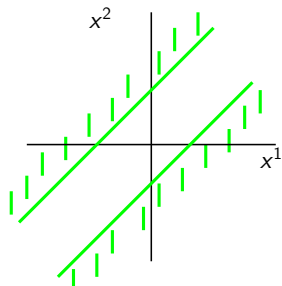
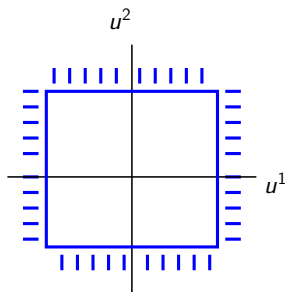
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- Constraints on states or outputs for reasons of safety, operability, product quality, etc.

$$Fx(k) \leq f \quad k \geq 0$$



Rate of change constraints

- Important in some applications to limit the *rate of change* of the input, $u(k) - u(k - 1)$

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- Augmented system model

$$\underbrace{\begin{bmatrix} x(k+1) \\ u(k) \end{bmatrix}}_{\tilde{x}(k+1)} = \underbrace{\begin{bmatrix} A & 0 \\ 0 & 0 \end{bmatrix}}_{\tilde{A}} \underbrace{\begin{bmatrix} x(k) \\ u(k-1) \end{bmatrix}}_{\tilde{x}(k)} + \underbrace{\begin{bmatrix} B \\ I \end{bmatrix}}_{\tilde{B}} u(k)$$

$$y(k) = \underbrace{\begin{bmatrix} C & 0 \end{bmatrix}}_{\tilde{C}} \underbrace{\begin{bmatrix} x(k) \\ u(k-1) \end{bmatrix}}_{\tilde{x}(k)}$$

Rate of change constraints

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is stated as

$$F\tilde{x}(k) + Eu(k) \leq e \quad F = \begin{bmatrix} 0 & -I \\ 0 & I \end{bmatrix} \quad E = \begin{bmatrix} I \\ -I \end{bmatrix} \quad e = \begin{bmatrix} \overline{\Delta} \\ -\underline{\Delta} \end{bmatrix}$$

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- General *linear* constraints

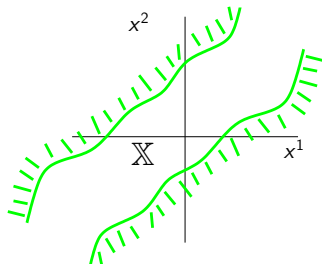
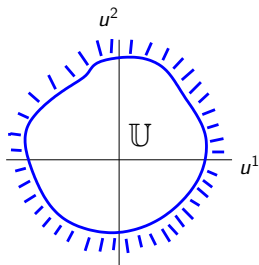
$$Fx(k) + Eu(k) \leq e \quad k \geq 0$$

subsumes all previous forms

Nonlinear constraints

- For nonlinear models, linear constraints are not required

$$u(k) \in \mathbb{U} \quad x(k) \in \mathbb{X} \quad k \geq 0$$



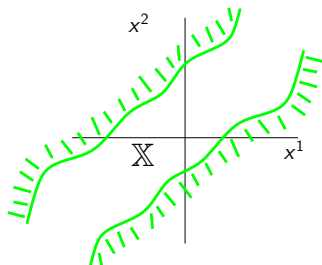
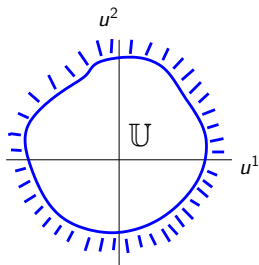
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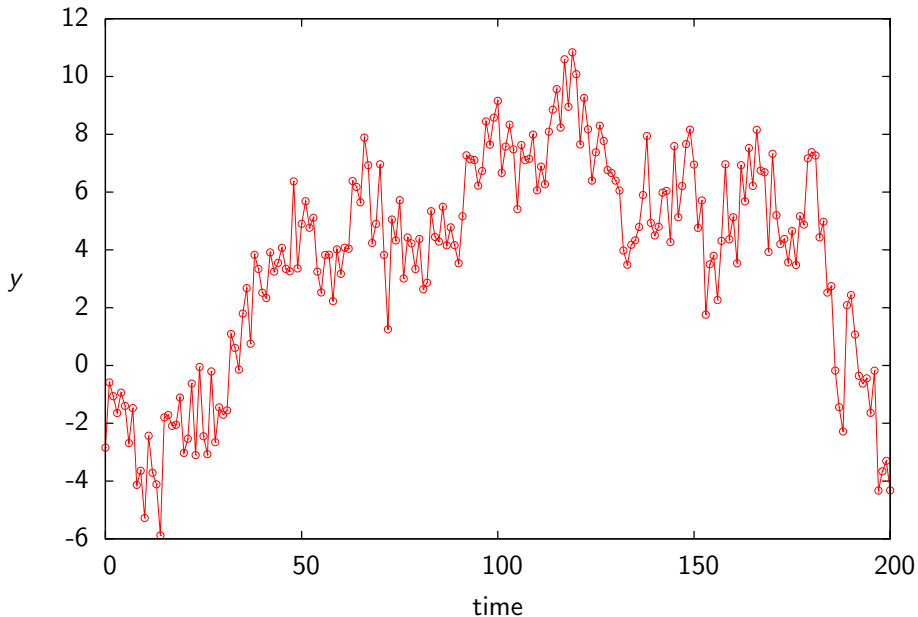
$$u(k) \in \mathbb{U} \quad x(k) \in \mathbb{X} \quad k \geq 0$$

- Even more general

$$(u(k), x(k)) \in \mathbb{Z} \quad k \geq 0$$



Typical process data



Random disturbances

$$\begin{aligned}x^+ &= Ax + Bu + Gw \\y &= Cx + v\end{aligned}$$

- w is the *random variable* affecting the process

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- w is the *random variable* affecting the process
- v is the *random variable* affecting the measurement

Random disturbances

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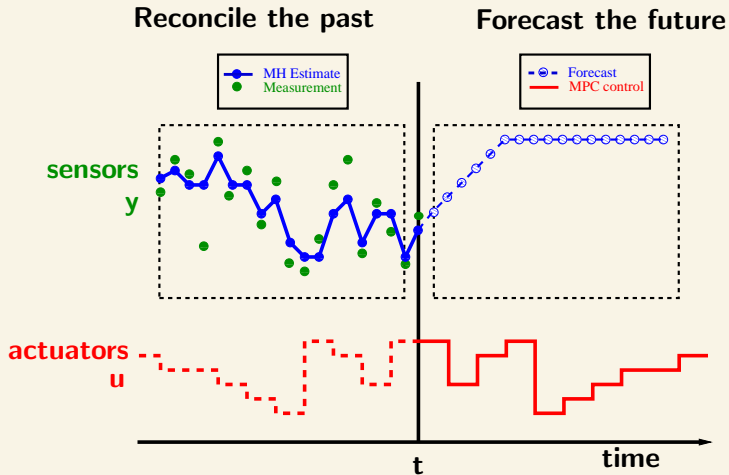
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Random disturbances

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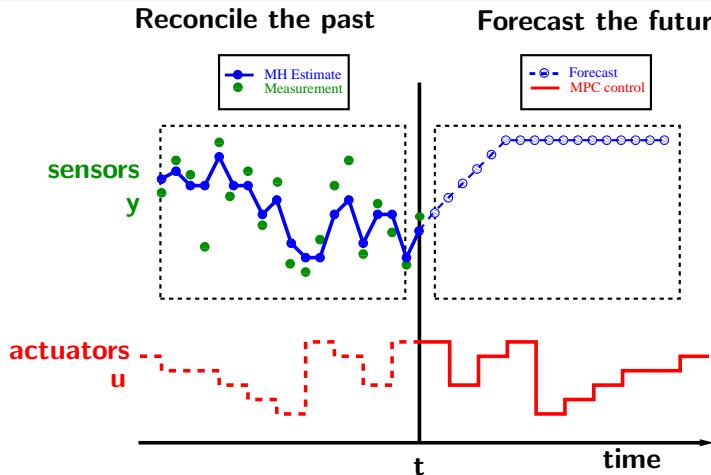
- w is the *random variable* affecting the process
- v is the *random variable* affecting the measurement
- So v models measurement noise and w models process disturbance
- The controller needs to know the *relative amounts of each disturbance*

The model predictive control framework

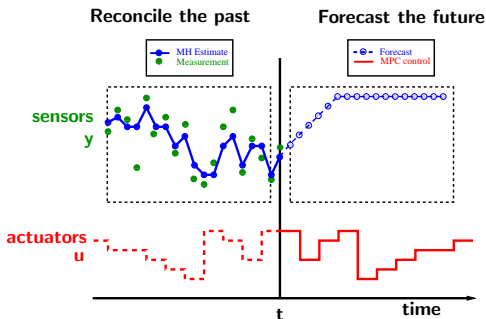


Predictive control

The future influences the present just as much as the past does.



Predictive control



$$\min_{u(t)} \int_0^T |y_{sp} - g(x, u)|_Q^2 + |u_{sp} - u|_R^2 dt$$

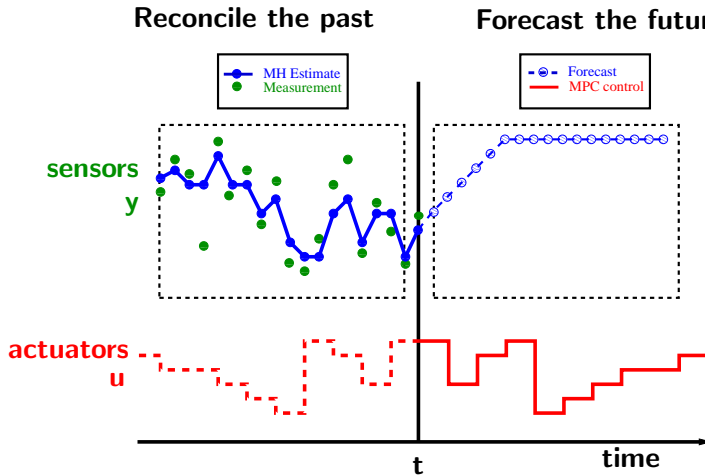
$$\dot{x} = f(x, u)$$

$$x(0) = x_0 \quad (\text{given})$$

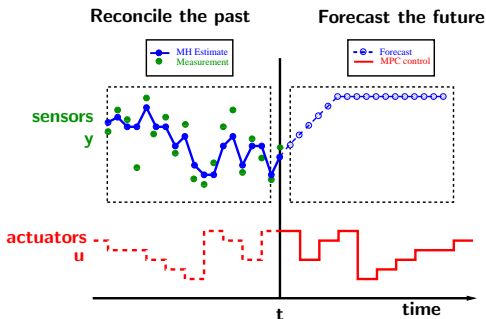
$$y = g(x, u)$$

State estimation

When I want to understand what is happening today or try to decide what will happen tomorrow, I look back.



State estimation

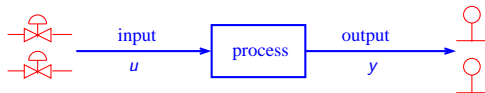


$$\min_{x_0, w(t)} \int_{-T}^0 |y - g(x, u)|_R^2 + |\dot{x} - f(x, u)|_Q^2 dt$$

$$\dot{x} = f(x, u) + w \quad (\text{process noise})$$

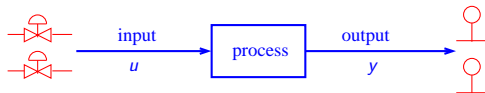
$$y = g(x, u) + v \quad (\text{measurement noise})$$

Separation of the control problem

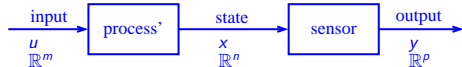


Input/output description

Separation of the control problem

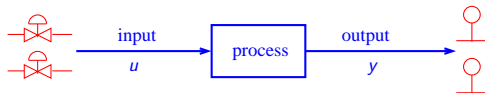


Input/output description

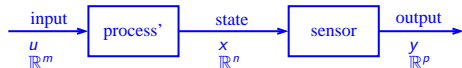


State description

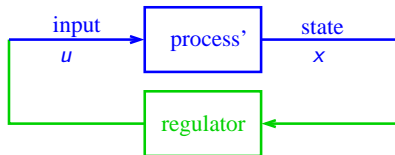
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Input/output description

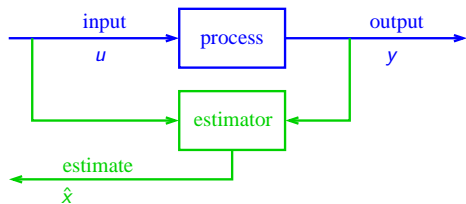


State description



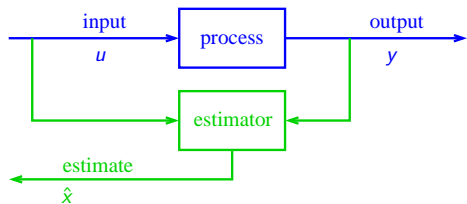
Regulation problem

Separation of the control problem

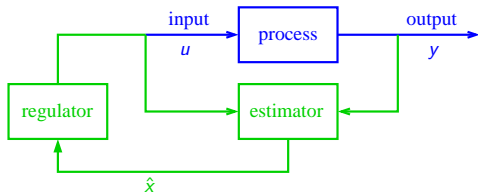


Estimation problem

Separation of the control problem



Estimation problem



Control problem

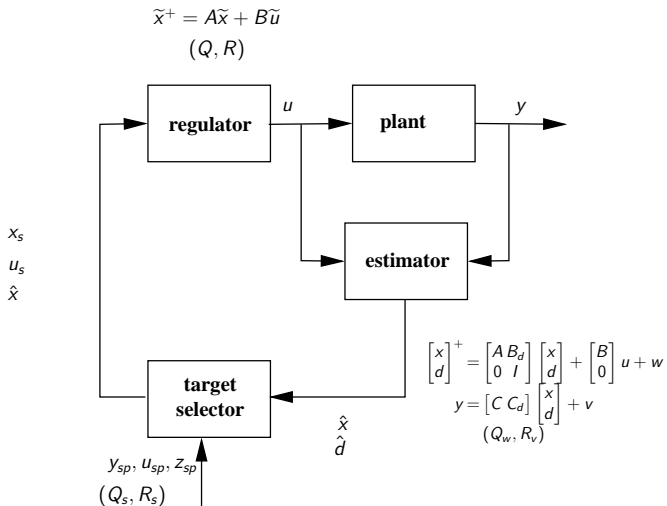
- Often want *zero tracking error* and rejection of *constant disturbances*

Controller Design for Zero Offset

- Often want *zero tracking error* and rejection of *constant disturbances*
- Augment the model with a constant disturbances (Davison and Smith, 1971, Kwakernaak and Sivan, 1972, p. 278) .

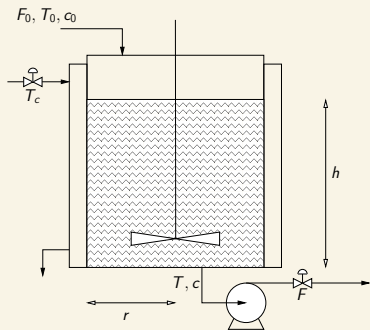
$$\begin{bmatrix} x(k+1) \\ d(k+1) \end{bmatrix} = \begin{bmatrix} A & B_d \\ 0 & I \end{bmatrix} \begin{bmatrix} x(k) \\ d(k) \end{bmatrix} + \begin{bmatrix} B \\ 0 \end{bmatrix} u(k) + w(k)$$
$$y(k) = [C \quad C_d] \begin{bmatrix} x(k) \\ d(k) \end{bmatrix} + v(k)$$

Offset free control system



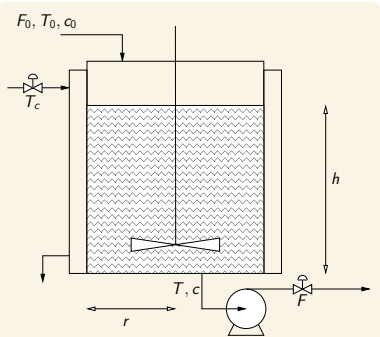
MPC controller consisting of: receding horizon regulator, state estimator, and target selector.

Design example. (Pannocchia and Rawlings, 2003)



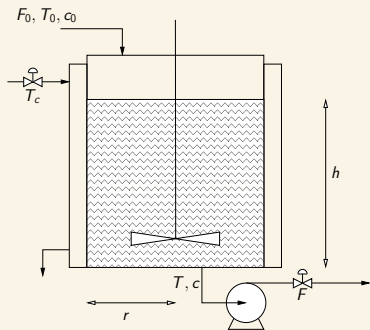
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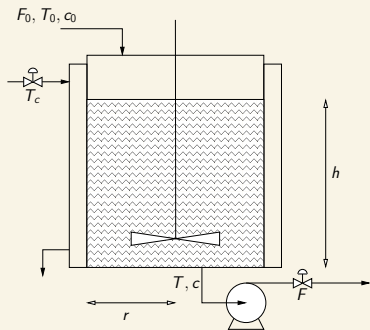
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- Controlled variables:
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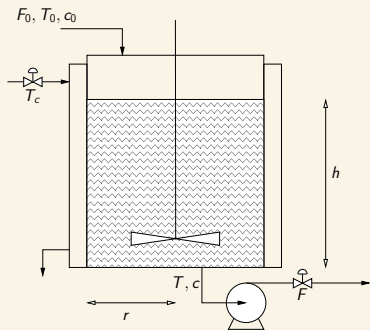
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- Controlled variables:
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 - ▶ concentration of species A (c)

Design example. (Pannocchia and Rawlings, 2003)



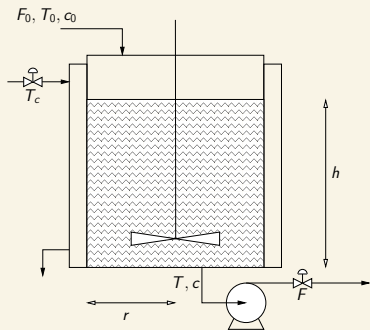
- Reaction $A \rightleftharpoons B$
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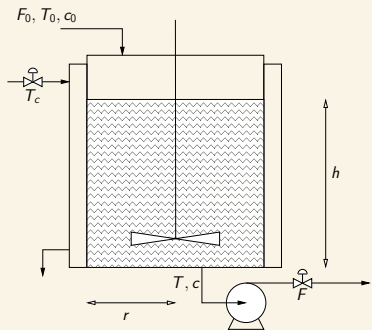
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- Component mass and energy balances

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- Nonlinear differential equation model:

$$\frac{dc}{dt} = \frac{F_0(c_0 - c)}{Ur^2h} - k_0c \exp\left(-\frac{E}{RT}\right)$$

$$\frac{dT}{dt} = \frac{F_0(T_0 - T)}{Ur^2h} + \frac{-\Delta H}{\rho C_p} k_0c \exp\left(-\frac{E}{RT}\right) + \frac{2U}{r\rho C_p}(T_c - T)$$

$$\frac{dh}{dt} = \frac{F_0 - F}{Ur^2}$$

Well-stirred reactor example: parameters

Parameter	Nominal value	Units
F_0	0.1	m^3/min
T_0	350	K
c_0	1	kmol/m^3
r	0.219	m
k_0	7.2×10^{10}	min^{-1}
E/R	8750	K
U	54.94	$\text{kJ}/\text{min} \cdot \text{m}^2 \cdot \text{K}$
ρ	1000	kg/m^3
C_p	0.239	$\text{kJ}/\text{kg} \cdot \text{K}$
ΔH	-5×10^4	kJ/kmol

Well-stirred reactor example: linearized model

- Using $\Delta = 1$ min, the linearized model is

$$\begin{aligned}x^+ &= Ax + Bu + B_p p \\ y &= Cx\end{aligned}$$

in which all the states are measured

$$x = \begin{bmatrix} c - c^s \\ T - T^s \\ h - h^s \end{bmatrix} \quad u = \begin{bmatrix} T_c - T_c^s \\ F - F^s \end{bmatrix} \quad y = \begin{bmatrix} c - c^s \\ T - T^s \\ h - h^s \end{bmatrix} \quad p = F_0 - F_0^s$$

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and

$$\begin{aligned}A &= \begin{bmatrix} 0.2681 & -0.00338 & -0.00728 \\ 9.703 & 0.3279 & -25.44 \\ 0 & 0 & 1 \end{bmatrix} & C &= \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix} \\ B &= \begin{bmatrix} -0.00537 & 0.1655 \\ 1.297 & 97.91 \\ 0 & -6.637 \end{bmatrix} & B_p &= \begin{bmatrix} -0.1175 \\ 69.74 \\ 6.637 \end{bmatrix}\end{aligned}$$

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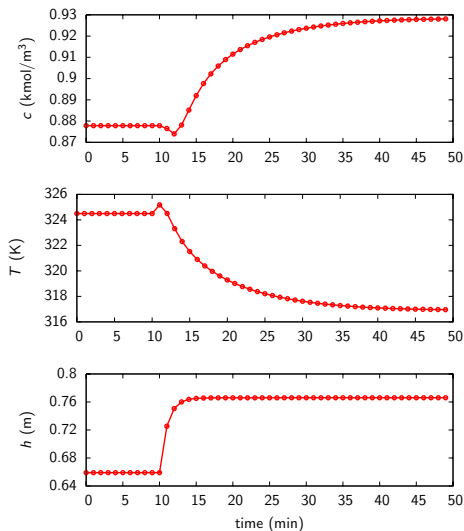
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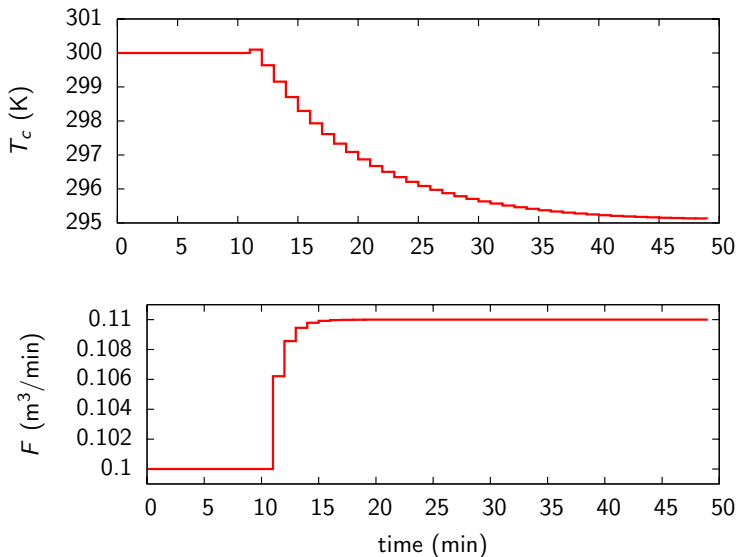
- Simulate the response of the controlled system
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- Is there steady offset in any of the outputs? Which ones?

Two integrating disturbances: output results



Three measured outputs versus time after a step change in F_0 at $t = 10$ min

Two integrating disturbances: input results



Two manipulated inputs versus time after a step change in F_0 at $t = 10$ min

Reactor case studies - two and three integrating disturbances

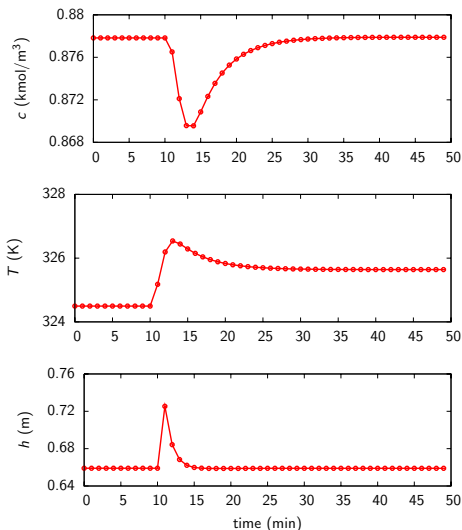
- Two integrating disturbances - conclusion:
 - ▶ Despite integrators in the two controlled variables, c and h , *all outputs have nonzero steady offset!*

Reactor case studies - two and three integrating disturbances

- Two integrating disturbances - conclusion:
 - ▶ Despite integrators in the two controlled variables, c and h , *all outputs have nonzero steady offset!*
- Next we try *three* integrating disturbances:
 - ▶ two added to the two controlled variables
 - ▶ one added to the second manipulated variable

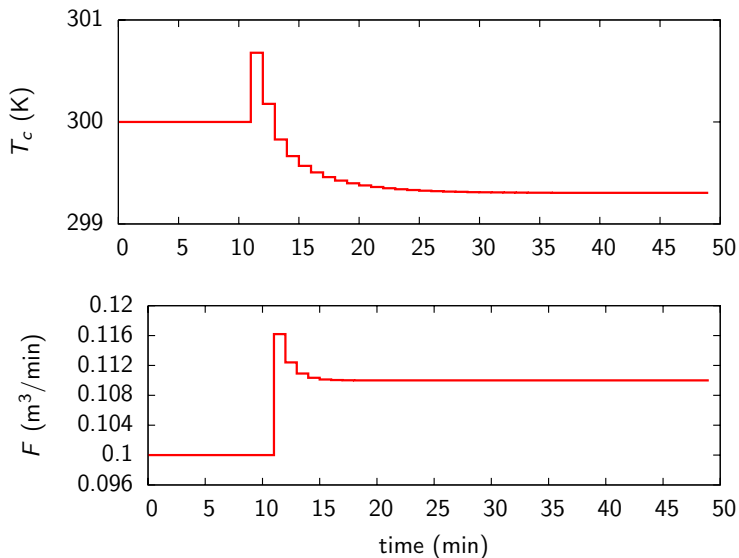
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Three integrating disturbances: output results



Three measured outputs versus time after a step change in F_0 at $t = 10$ min

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- With a more accurate disturbance model, better overall control is achieved
 - ▶ The controller uses smaller manipulated variable action and also achieves better output variable behavior
 - ▶ An added bonus is that steady offset is removed in the maximum possible number of outputs

Drug manufacturing versus petrochemicals — Differences

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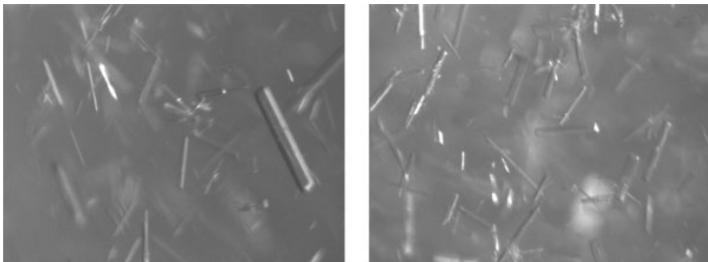
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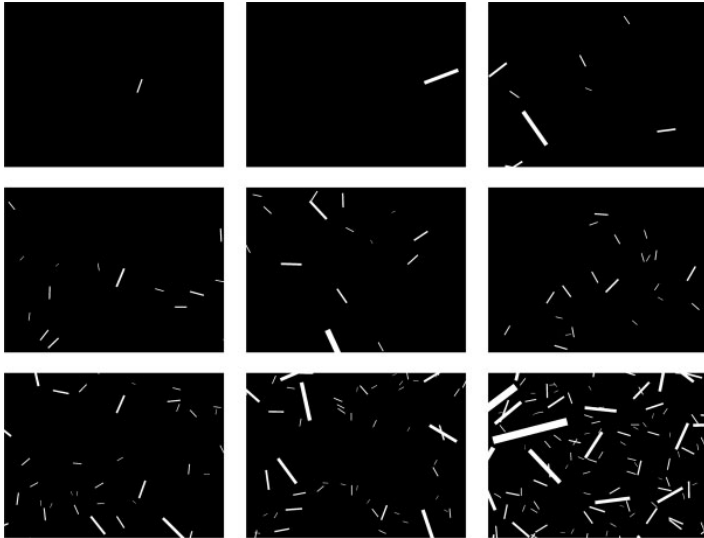
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- Small volume, high value-added products
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- Importance of the solid phase
- Novel solid-phase sensors (on-line video imaging)



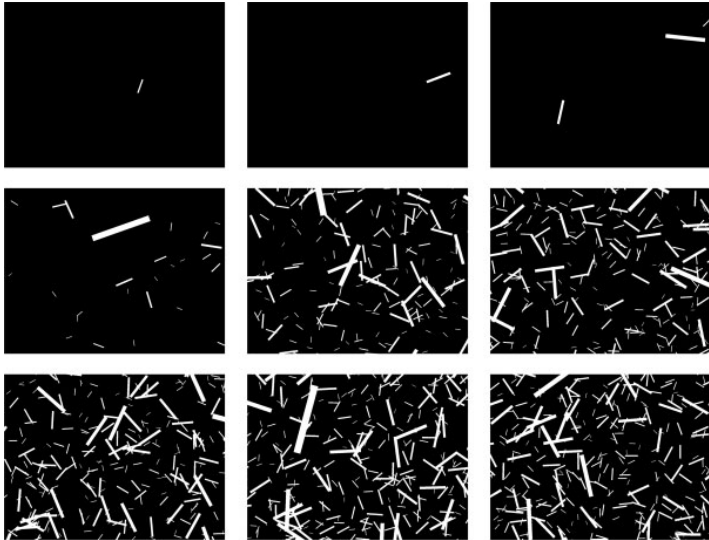
In situ images of rod-like crystals from our crystallization lab (Larsen and Rawlings, 2009).

Idealized data from a batch crystallization



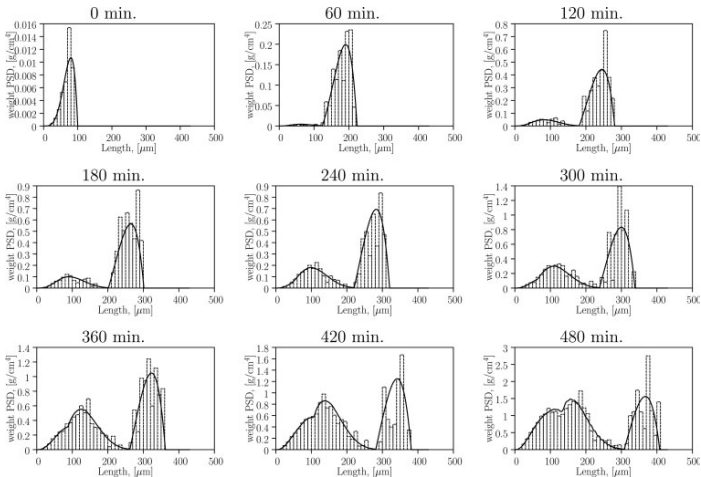
Optimal cooling profile. Images correspond to 60-minute intervals from 0 hours (upper left) to 8 hours (lower right).

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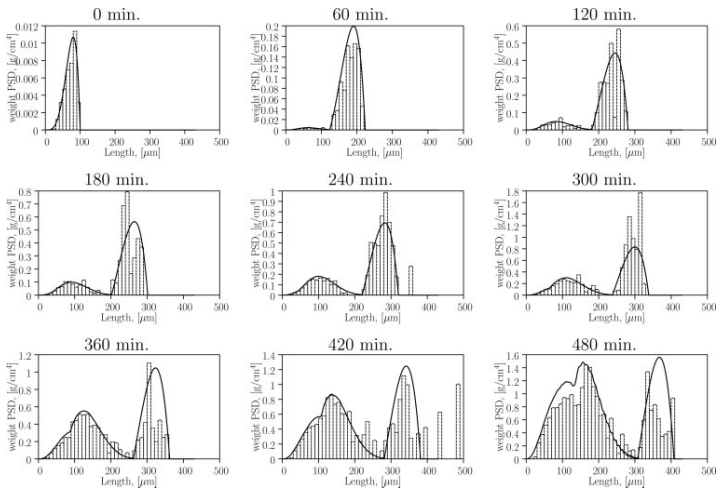
Linear cooling profile. Images correspond to 60-min intervals from 0 h (upper left) to 8 h (lower right).

Inferring size distribution from video images



Measured and estimated weight PSD for optimal cooling and perfect image analysis. Bin size = 10 μm and 100 bins.

Inferring size distribution from video images



Measured and estimated weight PSD for optimal cooling and image analysis using SHARC. Bin size = 10 μm and 100 bins.

Further reading I



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